# <u>Curriculum Topic</u> : Time-Harmonic Transmission Lines

## THT2 : Sinusoids on Transmission Lines

Module Outline:	
Prerequisite Skills	<u>Competencies</u>
Supplemental Reading and Resources	Assessments
Power Point Slides and Notes	

### **Prerequisite Skills**

Prerequisites / Requirements: THT1 Phasor Review

### **Competencies**

<b>Competency THT.2:</b>	Describe the mathematical form of sinusoidal waves on a
	transmission line

Competency Builders:

- THT.2.1 Write the voltage and current solutions for a time-harmonic transmission line in the phasor domain
- THT.2.2 Quantify the relationship between wavenumber, wavelength, and frequency

#### **Supplemental Reading and Resources**

Supplemental Reading Materials:

Prof. Peterson's online lecture notes 13

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### Assessments

The following questions and exercises may serve as either pre-assessment or postassessment tests to evaluate student knowledge.

 Question: THT.2.1
 Competency: THT.2.1

 True or False: Under steady-state time-harmonic excitation, a useful equivalent circuit for a transmission line is two short circuit pathways.

 Answer:

false

Question: THT.2.2 Competency: THT.2.2

A transmission line with velocity of propagation  $1.7 \times 10^8$  m/s is excited at 2.45 GHz. What is the wavenumber of this line?

Answer: 90.6 rad/m

Question: THT.2.3 Competency: THT.2.1

Transmission Line with Sinusoidal Excitation: (30 points) Below are phasor-domain voltage and current to a transmission line operating with steady-state sinusoids with frequency f:

$$\tilde{v}(z) = 100 \exp(-j\beta[z-D]) + 50 \exp(j\pi + j\beta[z-D])$$

$$\tilde{i}(z) = \exp(-j\beta[z-D]) - \frac{1}{2}\exp(j\pi + j\beta[z-D])$$

where z = 0 is the source side and z = D is the load side. Perform the following analysis:

- (a) In the equations above, circle the portion of the solution representing the backward-propagating current waveform. (5 points)
- (b) In the equations above, box the forward-propagating *amplitude* of the voltage waveform. (5 points)
- (c) Write a simplified expression for time-varying voltage evaluated at the front of the line (z = 0). Your answer should be in the time domain, v(t). (5 points)
- (d) What is the characteristic impedance,  $Z_0$ , of this line? (5 points)
- (e) What is the VSWR of voltages on this line? (5 points)
- (f) What is the load impedance,  $Z_L$ , for this line? (5 points)

Answer:

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- (a) In the equations above, circle the portion of the solution representing the backwardpropagating current waveform. The term  $\frac{1}{2} \exp(j\pi + j\beta[z - D])$  should be circled.
- (b) In the equations above, box the forward-propagating amplitude of the voltage waveform. The amplitude 100 should be boxed.
- (c) This follows from the phasor inverse transform definition (see formula sheet)

$$v(t,0) = \text{Real} \{ \tilde{v}(0) \exp(j2\pi ft) \}$$
  
= 100 cos(2\pi ft + \beta D) + 50 cos(2\pi ft - \beta D + \pi)  
= 100 cos(2\pi ft + \beta D) - 50 cos(2\pi ft - \beta D)

Anyone who made it to line two got full credit.

- (d)  $Z_0 = 100\Omega$
- (e) There are two ways of doing this. First, we immediately see that |Γ| = 0.5 because the reflected voltage/current wave is half the magnitude of the forward voltage/current wave. This can be plugged directly into one of the VSWR formulas to get VSWR=3. The same result can be obtained from the following reasoning: VSWR is the ratio of max voltage to min voltage on the line; the max voltage occurs when the forward wave adds in phase with the backwards wave (100+50 V); the min voltage occurs when the forward wave adds out-of-phase (destructively) with the backwards wave (100-50 V); thus, the VSWR will be 150/50 or 3.
- (f) For this problem, I gave full credit to anyone who put down line 1 of the following answer, but did not finish the calculation. Very few people got full credit for this problem.

$$Z_L = \frac{\tilde{v}(D)}{\tilde{i}(D)}$$

$$= \frac{100 + 50 \exp(j\pi)}{1 - \frac{1}{2}\exp(j\pi)}$$

$$= \frac{100 - 50}{1 + \frac{1}{2}}$$

$$= \frac{100}{3}\Omega$$

Many people tried to solve this from the VSWR by using the property  $|\Gamma| = \frac{1}{2}$  to backsolve  $\Gamma = \frac{Z_L - Z_0}{Z_L + Z_0}$ . This actually doesn't work because it only gives you the *magnitude* of the reflection coefficient. Many people worked the problem with  $\Gamma = \frac{1}{2}$  and found  $Z_L = 300\Omega$ , when in fact the reflection coefficient for this particular example was  $\Gamma = -\frac{1}{2}$ .

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